

# Quantum noise measurements: challenges and tricks

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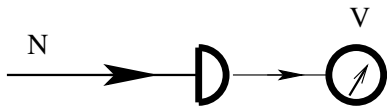
The College of William & Mary

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# Outline

# Quantum noise

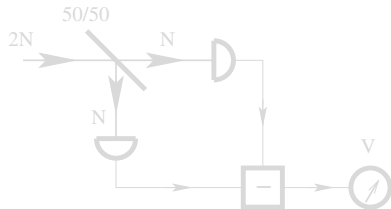
## Simple photodetector



$$V \sim N$$

$$\Delta V \sim \sqrt{N}$$

## Balanced photodetector

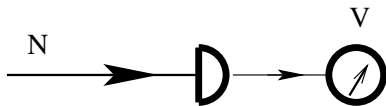


$$V = 0$$

$$\Delta V \sim \sqrt{N}$$

# Quantum noise

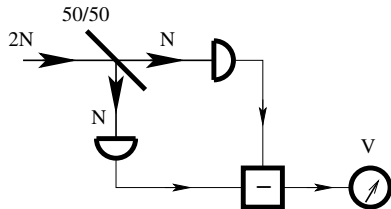
## Simple photodetector



$$V \sim N$$

$$\Delta V \sim \sqrt{N}$$

## Balanced photodetector



$$V = 0$$

$$\Delta V \sim \sqrt{N}$$

# Coherent states of light: $\alpha$

## Quantum Mechanics

### Properties:

- $\bar{N} = \langle \alpha | \hat{N} | \alpha \rangle = \langle \alpha | \mathbf{a}^\dagger \mathbf{a} | \alpha \rangle = \alpha^2$
- $\Delta N = \sqrt{\langle \alpha | \hat{N}^2 - \bar{N}^2 | \alpha \rangle} = \alpha$

If  $\alpha \rightarrow 0$

- $\Delta N = 1/2$

### Classically

- Electric field  $E = E_0 + \delta E$
- Intensity  $I = E^2 = (E_0 + \delta E)^2 = E_0^2 + 2E_0\delta E + \delta E^2 \approx I_0 + 2\sqrt{I_0}\delta E$ 
  - $\bar{I} = I_0$
  - $\Delta I = 2\sqrt{I_0}\delta E$

- In ideal world each photon generate an electron on a photodiode
- Photodiode measures intensity or photon number
- Photodiode output current ( $i$ ) proportional to intensity ( $I$ )

So

$$i_o = \eta e \frac{I_0 \text{Area}}{\hbar\omega} = \eta e \frac{P}{\hbar\omega} = \eta 0.64 [A/W] P$$

Shot-noise

$$\Delta i \sim \Delta I = \sqrt{2ei_oRBW}$$

# More about Shot Noise

$$i_o = \eta 0.64 [A/W] P$$

$$\Delta i \sim \Delta I = \sqrt{2 e i_o R B W}$$

Typical numbers:

- $P = 1 mW$
- $\eta = 100 \%$
- $i_o = .64 mA$

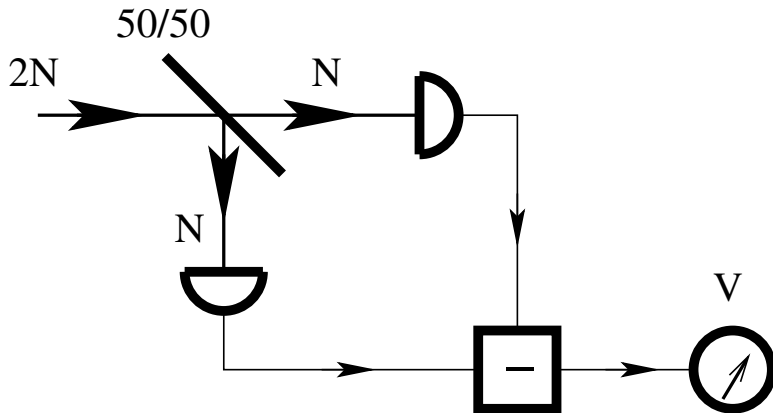
$$\Delta i = 14 \text{ pA} / \sqrt{Hz}$$

- Typical Scope RBW=20 MHz,  $\Delta i = 64 \text{ nA}$
- We are aiming to RBW=100kHz,  $\Delta i = 4.5 \text{ nA}$

We are trying to resolve tenth of the inch on the top of the Empire State Building.

# Measurement: step 1 kill DC current offset

Relatively easy with balanced photodetector



Watch out

- Matched quantum efficiency
- We need real 50/50 beam splitter



# Measurement: step 2 transform current to voltage

Very simple: Load resistor



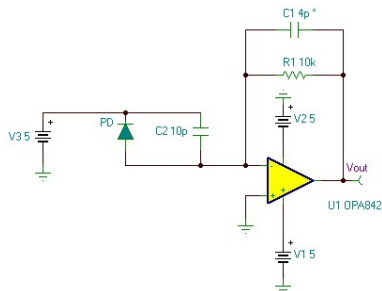
- Cut off frequency is still  $f_c = 1/(RC_d)$
- Current transimpedance gain  $\frac{R_f}{1+jR_fC_d\omega}$ ,

But  $C_d$  goes up as  $V$  goes up

- the more gain we have the weaker response at high frequencies

# Measurement: step 2 improved current to voltage

Very simple: Load resistor

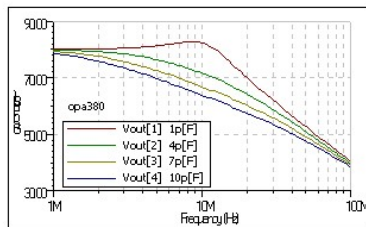
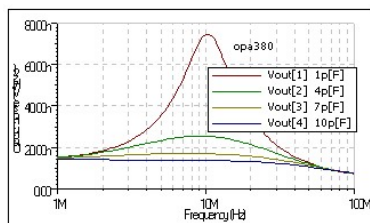


- Cut off frequency is still  $f_c = 1/(RC_d)$
- Current transimpedance gain  $\frac{R_f}{1+jR_fC_d\omega}$ ,
- $C_d = \text{const}$

# Problems with amplifier

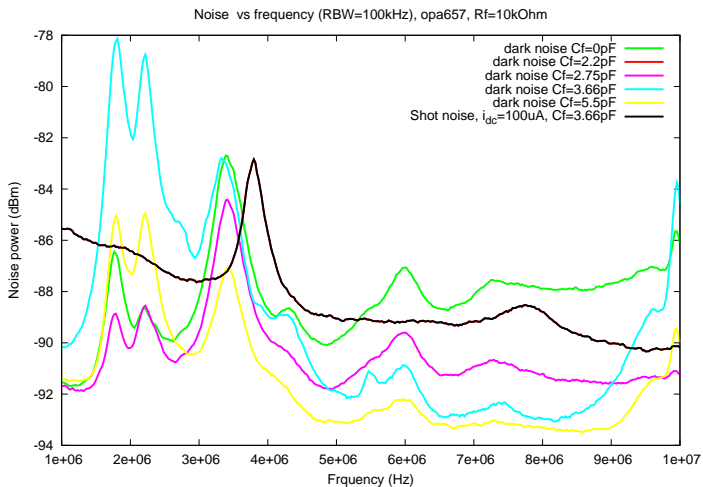
- Current transimpedance gain  $\frac{R_f}{1+jR_fC_d\omega}$ , drops with frequency
- Voltage gain =  $\frac{Z_f}{Z_{in}} = jR_fC_d\omega$ , grows with frequency

**We need compensation:** introducing  $C_f$

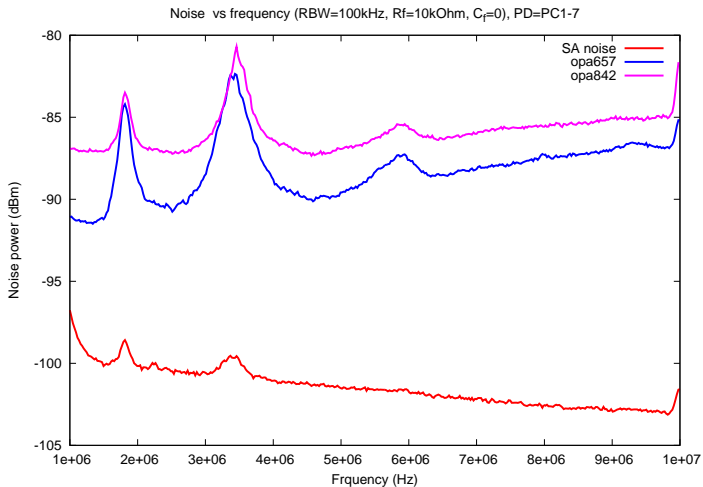


There is a price to pay: we decrease cut off frequency

# Noise suppression with $C_F$



# Dark noise with different OpAmps



# Summary

- We leave in the fallen word full of noise
- Measuring quantum noise is hard
- Be prepared to scarify either SNR or bandwidth
- Careful choice of components is required

## Good news

- We have 3-4 dB separation between quantum noise (our signal) and electronic noise

## Bad news

- Perfect PD is still to be found, since QE is not great, just 84%